## Features

- 36 V input VI Chip ${ }^{\oplus}$ PRM
- Vin range 18 - 60 Vdc
- High density - $407 \mathrm{~W} / \mathrm{in}^{3}$
- Small footprint - 1.1 in $^{2}$
- Low weight - 0.5 oz ( 15 g )
- Adaptive Loop feedback
- ZVS buck-boost regulator
- 1.35 MHz switching frequency
- 95\% efficiency
- $125^{\circ} \mathrm{C}$ operation ( T j )


$$
\begin{aligned}
& \text { Vin }=18-60 \mathrm{Vdc} \\
& \mathrm{~V}_{\mathrm{f}}=26-55 \mathrm{Vdc} \\
& \mathrm{P}_{\mathrm{f}}=120 \mathrm{~W} \\
& \mathrm{If}_{\mathrm{f}}=2.5 \mathrm{~A}
\end{aligned}
$$

## Product Description

The VI Chip regulator is a very efficient non-isolated regulator capable of both boosting and bucking a wide range input voltage. It is specifically designed to provide a controlled Factorized Bus distribution voltage for powering downstream VTM Transformer - fast, efficient, isolated, low noise Point-of-Load (POL) converters. In combination, PRMs and VTMs form a complete DC-DC converter subsystem offering all of the unique benefits of Vicor's Factorized Power Architecture ${ }^{\text {TM }}$ (FPA) ${ }^{\text {TM }}$ : high density and efficiency; low noise operation; architectural flexibility; extremely fast transient response; and elimination of bulk capacitance at the Point-of-Load (POL).
In FPA systems, the POL voltage is the product of the Factorized Bus voltage delivered by the PRM and the "K-factor" (the fixed voltage transformation ratio) of a downstream VTM. The PRM controls the Factorized Bus voltage to provide regulation at the POL. Because VTMs perform true voltage division and current multiplication, the Factorized Bus voltage may be set to a value that is substantially higher than the bus voltages typically found in "intermediate bus" systems, reducing distribution losses and enabling use of narrower distribution bus traces. A PRM-VTM chip set can provide up to 100 A , or 115 W at a FPA system density of $169 \mathrm{~A} / \mathrm{in}^{3}$, or $195 \mathrm{~W} / \mathrm{in}^{3}$ - and because the PRM can be located, or "factorized, " remotely from the POL, these power densities can be effectively doubled.
The PRM described in this data sheet features a unique "Adaptive Loop" compensation feedback: a single wire alternative to traditional remote sensing and feedback loops that enables precise control of an isolated POL voltage without the need for either a direct connection to the load or for noise sensitive, bandwidth limiting, isolation devices in the feedback path.

## Absolute Maximum Ratings

| Parameter | Values | Unit | Notes |
| :---: | :---: | :---: | :---: |
| +ln to -In | -1.0 to 85.0 | Vdc |  |
| PC to -In | -0.3 to 6.0 | Vdc |  |
| PR to -In | -0.3 to 9.0 | Vdc |  |
| IL to -In | -0.3 to 6.0 | Vdc |  |
| VC to -In | -0.3 to 18.0 | Vdc |  |
| +Out to -Out | -0.3 to 59 | Vdc |  |
| SC to -Out | -0.3 to 3.0 | Vdc |  |
| VH to -Out | -0.3 to 9.5 | Vdc |  |
| OS to -Out | -0.3 to 9.0 | Vdc |  |
| CD to -Out | -0.3 to 9.0 | Vdc |  |
| SG to -Out | 100 | mA |  |
| Continuous output current | 2.5 | Adc |  |
| Continuous output power | 120 | W |  |
| Case temperature during reflow | 245 | ${ }^{\circ} \mathrm{C}$ | MSL 4 <br> (Datecode 1528 and later) |
| Operating junction temperature | -40 to 125 | ${ }^{\circ} \mathrm{C}$ | T-Grade |
| Storage temperature | -40 to 125 | ${ }^{\circ} \mathrm{C}$ | T-Grade |

## DC-DC Converter



The P036F048T12AL is used with any 048 input series VTM to provide a regulated and isolated output.

## General Specifications

Part Numbering


## Overview of Adaptive Loop Compensation

Adaptive Loop compensation, illustrated in Figure 1, contributes to the bandwidth and speed advantage of Factorized Power. The PRM monitors its output current and automatically adjusts its output voltage to compensate for the voltage drop in the output resistance of the VTM. Ros sets the desired value of the VTM output voltage, Vout; R $\mathrm{R}_{\mathrm{CD}}$ is set to a value that compensates for the output resistance of the VTM (which, ideally, is located at the point of load). For selection of Ros and $R_{C D}$, refer to Table 1 below or Page 9.

The VI Chip's bi-directional VC port :

1. Provides a wake up signal from the PRM to the VTM that synchronizes the rise of the VTM output voltage to that of the PRM.
2. Provides feedback from the VTM to the PRM to enable the PRM to compensate for the voltage drop in VTM output resistance, $\mathrm{R}_{\mathrm{O}}$.


Figure 1 - With Adaptive Loop control, the output of the VTM is regulated over the load current range with only a single interconnect between the PRM and VTM and without the need for isolation in the feedback path.

| Desired Load Voltage (Vdc) | VTM P/N(1) | Max VTM Output Current (A) ${ }^{(2)}$ | $\mathrm{R}_{\text {OS }}(\mathrm{k} \Omega)^{(3)}$ | $\mathbf{R}_{\text {CD }}(\Omega){ }^{(3)}$ |
| :---: | :---: | :---: | :---: | :---: |
| 1.0 | V048F015T100 | 100 | 3.57 | 26.1 |
| 1.2 | V048F015T100 | 100 | 2.94 | 32.4 |
| 1.5 | V048F015T100 | 100 | 2.37 | 39.2 |
| 1.8 | V048F020T080 | 80 | 2.61 | 35.7 |
| 2.0 | V048F020T080 | 80 | 2.37 | 39.2 |
| 3.0 | V048F030T070 | 70 | 2.37 | 39.2 |
| 3.3 | V048F040T050 | 50 | 2.89 | 32.6 |
| 5.0 | V048F060T040 | 40 | 2.87 | 33.2 |
| 8.0 | V048F080T030 | 30 | 2.37 | 32.9 |
| 9.6 | V048F096T025 | 25 | 2.37 | 32.9 |
| 10 | V048F120T025 | 25 | 2.86 | 32.9 |
| 12 | V048F120T025 | 25 | 2.37 | 39.2 |
| 15 | V048F160T015 | 15 | 2.49 | 37.4 |
| 24 | V048F240T012 | 12.5 | 2.37 | 39.2 |
| 28 | V048F320T009 | 9.4 | 2.74 | 35.7 |
| 36 | V048F480T006 | 6.3 | 3.16 | 30.1 |
| 48 | V048F480T006 | 6.3 | 2.37 | 39.2 |

## Note:

(1) See Table 2 on page 9 for nominal Vout range and K factors.
(2) See "PRM output power vs. VTM output power" on Page 10
(3) $1 \%$ precision resistors recommended

Table 1 - Configure your Chip Set using the PRM-AL

## Electrical Specifications

Input Specs (Conditions are at $36 \mathrm{Vin}, 48 \mathrm{Vf}$, full load, and $25^{\circ} \mathrm{C}$ ambient unless otherwise specified)

| Parameter | Min | Typ | Max | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :--- |
| Input voltage range | 18 | 36 | 60 | Vdc |  |
| Input dV/dt |  |  | 1 | $\mathrm{~V} / \mu \mathrm{s}$ |  |
| Input undervoltage turn-on | 15 | 17.6 | Vdc |  |  |
| Input undervoltage turn-off | 60 | 15.9 |  | Vdc |  |
| Input overvoltage turn-on | 62 |  | Vdc |  |  |
| Input overvoltage turn-off | 63 | 65 | Vdc |  |  |
| Input quiescent current | 0.5 | 1 | mA | PC low |  |
| Input current | 3.5 |  | Adc |  |  |
| Input reflected ripple current | 586 |  | $\mathrm{~mA} \mathrm{p-p}$ | See Figures 4 \& 5 |  |
| No load power dissipation | 3 | 6 | W |  |  |
| Internal input capacitance | 5 |  | $\mu \mathrm{~F}$ | Ceramic |  |
| Recommended external input capacitance | 100 |  | $\mu \mathrm{~F}$ | See Figure 5 for input filter circuit. <br> Source impedance dependent |  |

## Input Waveforms



Figure 2 -Vf and $P C$ response from power up


Figure 4 - Input reflected ripple current


Figure 3 - Vf turn-on waveform with inrush current - PC enabled


Figure 5 - Input filter capacitor recommendation

## Electrical Specifications (continued)

Output Specs (Conditions are at $36 \mathrm{Vin}, 48 \mathrm{Vf}$, full load, and $25^{\circ} \mathrm{C}$ ambient unless otherwise specified)

| Parameter | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Output voltage range | 26 | 48 | 55 | Vdc | Factorized Bus voltage (Vf) set by Ros |
| Output power | 0 |  | 120 | W |  |
| Output current | 0 |  | 2.5 | Adc |  |
| DC current limit | 2.6 | 2.96 | 3.3 | Adc | $I_{\llcorner }$pin floating |
| Average short circuit current |  |  | 1.25 | A | Auto recovery |
| Set point accuracy |  | 1.5 |  | \% |  |
| Line regulation |  | 0.1 | 0.2 | \% | Low line to high line |
| Load regulation |  | 0.1 | 0.2 | \% | No CD resistor |
| Load regulation (at VTM output) |  | 1.0 | 2.0 | \% | Adaptive Loop |
| Current share accuracy |  | 5 | 10 | \% | See description page 8 |
| Efficiency |  |  |  |  |  |
| Full load |  | 95 |  | \% | See Figure 6,7 \& 8 |
| Output overvoltage set point | 56 |  | 59.4 | Vdc |  |
| Output ripple voltage |  |  |  |  |  |
| No external bypass |  | 1.51 | 3.5 | \% | Factorized Bus, see Figure 13 |
| With $10 \mu \mathrm{~F}$ capacitor |  | 0.42 | 1.0 | \% | Factorized Bus, see Figure 14 |
| Switching frequency | 1.26 | 1.35 | 1.46 | MHz | Fixed frequency |
| Output turn-on delay |  |  |  |  |  |
| From application of power |  | 74 | 250 | ms | See Figure 2 |
| From PC pin high |  | 100 |  | $\mu \mathrm{s}$ | See Figure 3 |
| Internal output capacitance |  | 5 |  | $\mu \mathrm{F}$ | Ceramic |
| Factorized Bus capacitance |  |  | 47 | $\mu \mathrm{F}$ |  |

## Efficiency Graphs



Figure 6 - Efficiency vs. output current at 48 Vf


Figure 8 - Efficiency vs. output current at 26 Vf


Figure 7 - Efficiency vs. output current at 36 Vf

## Electrical Specifications (continued)

VI Chip Regulator

## Output Waveforms



Figure 9 - Transient response; PRM alone 36 Vin, 0-2.5-0 A no load capacitance, local loop and 48 Vf


Figure 11 - Transient response; PRM alone 60 Vin, 0-2.5-0 A no load capacitance, local loop and 48 Vf


Figure 13 - Output ripple full load no bypass capacitance. $V f=48 \mathrm{Vdc}$


Figure 10 - Transient response; PRM alone 18 Vin, 0-2.5-0 A no load capacitance, local loop and 48 Vf


Figure 12 - PC during fault - frequency will vary as a function of line voltage


Figure 14 - Output ripple full load $10 \mu \mathrm{~F}$ bypass capacitance. $V f=48 \mathrm{Vdc}$

Auxiliary Pins (Conditions are at $36 \mathrm{Vin}, 48 \mathrm{Vf}$, full load, and $25^{\circ} \mathrm{C}$ ambient unless otherwise specified)

| Parameter | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: |
| VC (VTM Control) |  |  |  |  |  |
| Pulse width | 8 | 12 | 18 | ms |  |
| Peak voltage | 12 | 14 | 18 | V | Referenced to -Out |
| PC (Primary Control) |  |  |  |  |  |
| DC voltage | 4.8 | 5.0 | 5.2 | Vdc | Referenced to -In |
| Module disable voltage | 2.3 | 2.4 |  | Vdc | Referenced to -In |
| Module enable voltage |  | 2.5 | 2.6 | Vdc |  |
| Disable hysteresis |  | 100 |  | mV |  |
| Current limit |  | 1.75 | 1.90 | mA | Source only after start up; not to be used for aux. supply; $100 \mathrm{k} \Omega$ minimum load impedance to assure start up. |
| Enable delay time |  | 100 |  | $\mu \mathrm{s}$ |  |
| Disable delay time |  | 1 |  | $\mu \mathrm{s}$ |  |
| IL (Current Limit Adjust) |  |  |  |  |  |
| Voltage |  | 1 |  | V |  |
| Accuracy |  | $\pm 15$ |  | \% | Based on DC current limit set point |
| PR (Parallel Port) |  |  |  |  |  |
| Voltage | 0.5 |  | 3.5 | V | Referenced to SG; See description page 8 |
| Source current | 1 |  |  | mA |  |
| External capacitance |  |  | 100 | pF |  |
| VH (Auxiliary Voltage) |  |  |  |  | Typical internal bypass $\mathrm{C}=0.1 \mu \mathrm{~F}$ |
| Range | 8.7 | 9.0 | 9.3 | Vdc | Maximum external $\mathrm{C}=0.1 \mu \mathrm{~F}$, referenced to SG |
| Regulation |  | 0.04 |  | \%/mA |  |
| Current |  |  | 5 | mAp | See description page 8 |
| SC (Secondary Control) |  |  |  |  |  |
| Voltage | 1.23 | 1.24 | 1.25 | Vdc | Referenced to SG |
| Internal capacitance |  | 0.22 |  | $\mu \mathrm{F}$ |  |
| External capacitance |  |  | 0.7 | $\mu \mathrm{F}$ |  |
| OS (Output Set) |  |  |  |  |  |
| Set point accuracy |  | $\pm 1.5$ |  | \% | Includes 1\% external resistor |
| Reference offset |  | $\pm 4$ |  | mV |  |
| CD (Compensation Device) |  |  |  |  |  |
| External resistance | 20 |  |  | $\Omega$ | Omit resistor for regulation at output of PRM |

## General Specs

| Parameter | Min | Typ | Max | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: |
| MTBF |  |  |  |  |  |
| MIL-HDBK-217F |  | 2.2 |  | Mhrs | $25^{\circ} \mathrm{C}$, GB |
| Agency approvals |  | cTÜVus |  |  | UL/CSA 60950-1, EN60950-1 |
|  |  | CE Marked for Low Voltage Directive and RoHS Recast Directive, as applicable |  |  |  |
| Mechanical parameters |  |  |  |  | See Mechanical Drawings, Figures 19-22 |
| Weight |  | 0.53/15 |  | oz/g |  |
| Dimensions |  |  |  |  |  |
| Length |  | 1.28/32,5 |  | $\mathrm{in} / \mathrm{mm}$ |  |
| Width |  | 0.87/22,0 |  | $\mathrm{in} / \mathrm{mm}$ |  |
| Height |  | 0.265/6,73 |  | $\mathrm{in} / \mathrm{mm}$ |  |
| Thermal |  |  |  |  |  |
| Over temperature shutdown | 125 | 135 | 140 | ${ }^{\circ} \mathrm{C}$ | Junction temperature |
| Thermal capacity |  | 9.3 |  | Ws/ ${ }^{\circ} \mathrm{C}$ |  |
| Junction-to-case thermal impedance ( $\mathrm{R}_{\text {өJ¢ }}$ ) |  | 1.1 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |  |
| Junction-to-board thermal impedance ( $\mathrm{R}_{\text {өBB }}$ ) |  | 2.1 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |  |
| Case-to-ambient |  | 3.7 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ | With 0.25" heat sink @ 300 LFM |

## Pin / Control Functions

## +In / -In DC Voltage Ports

The VI Chip maximum input voltage should not be exceeded. PRMs have internal over / undervoltage lockout functions that prevent operation outside of the specified input range. PRMs will turn on when the input voltage rises above its undervoltage lockout. If the input voltage exceeds the overvoltage lockout, PRMs will shut down until the overvoltage fault clears. PC will toggle indicating an out of bounds condition.

## +Out / -Out Factorized Voltage Output Ports

These ports provide the Factorized Bus voltage output. The -Out port is connected internally to the -In port through a current sense resistor. The PRM has a maximum power and a maximum current rating and is protected if either rating is exceeded. Do not short -Out to -In.

## VC - VTM Control

The VTM Control (VC) port supplies an initial $\mathrm{V}_{\mathrm{CC}}$ voltage to downstream VTMs, enabling the VTMs and synchronizing the rise of the VTM output voltage to that of the PRM. The VC port also provides feedback to the PRM to compensate for voltage drop due to the VTM output resistance. The PRM's VC port should be connected to the VTM VC port. A PRM VC port can drive a maximum of two (2) VTM VC ports.

## PC - Primary Control

The PRM voltage output is enabled when the PC pin is open circuit (floating). To disable the PRM output voltage, the PC pin is pulled low. Open collector optocouplers, transistors, or relays can be used to control the PC pin. When using multiple PRMs in a high power array, the PC ports should be tied together to synchronize their turn on. During an abnormal condition the PC pin will pulse (Fig.12) as the PRM initiates a restart cycle. This will continue until the abnormal condition is rectified. The PC should not be used as an auxiliary voltage supply, nor should it be switched at a rate greater than 1 Hz .

## TM - Factory Use Only

## IL - Current Limit Adjust

The PRM has a preset, maximum, current limit set point. The IL port may be used to reduce the current limit set point to a lower value. See "adjusting current limit" on page 10.

## PR - Parallel Port

The PR port signal, which is proportional to the PRM output power, supports current sharing of two PRMs. To enable current sharing, PR ports should be interconnected. Steps should be taken to minimize coupling noise into the interconnecting bus. Terminate this port with a 10 k equivalent resistance to SG, e.g. 10 k for a single PRM, 20 k each for 2 PRMs in parallel, 30 k each for 3 PRMs in parallel etc.. Please consult Vicor Applications Engineering regarding additional considerations when paralleling more than two PRMs.

## VH - Auxiliary Voltage

VH is a gated (e.g. mirrors PC), non-isolated, nominally 9 Volt, regulated DC voltage (see "Auxiliary Pins" specifications, on Page 7) that is referenced to SG. VH may be used to power external circuitry having a total current consumption of no more than 5 mA under either transient or steady state conditions including turn-on.

## Application Information



Figure 16 - Adaptive Loop compensation with soft start using the SC port.

## Output Voltage Setting with Adaptive Loop

The equations for calculating Ros and $R_{C D}$ to set a VTM output voltage are:
$R_{\text {OS }}=\frac{93100}{\left(\frac{\mathrm{~V}_{\mathrm{L}} \cdot 0.8395}{\mathrm{~K}}\right)-1}$
$R_{C D}=\quad \frac{91238}{R_{O S}}+1$
(2)
$V_{L}=$ Desired load voltage
$\mathrm{V}_{\text {OUT }}=$ VTM output voltage
$\mathrm{K}=$ VTM transformation ratio
(available from appropriate VTM data sheet)
$V_{f}=$ PRM output voltage, the Factorized Bus (see Figure 16)
$R_{0}=$ VTM output resistance
(available from appropriate VTM data sheet)
$L_{L}=$ Load Current
(actual current delivered to the load)

## Output Voltage Trimming (optional)

After setting the output voltage from the procedure above the output may be margined down ( 26 Vf min ) by a resistor from SC-SG using this formula:
$\mathrm{R}_{\mathrm{d}} \Omega=\frac{10000 \mathrm{~V}_{\mathrm{fd}}}{\mathrm{V}_{\mathrm{fs}}-\mathrm{V}_{\mathrm{fd}}}$

Where $\mathrm{V}_{\mathrm{fd}}$ is the desired factorized bus and $\mathrm{V}_{\mathrm{fs}}$ is the set factorized bus.
A low voltage source can be applied to the SC port to margin the load voltage in proportion to the SC reference voltage.
An external capacitor can be added to the SC port as shown in Figure 16 to control the output voltage slew rate for soft start.

| Nominal Vout <br> Range (Vdc) | VTM <br> K Factor |  |
| :---: | :---: | :---: |
| $0.8 \leftrightarrow y 1.6$ | $1 / 32$ |  |
| $1.1 \leftrightarrow 2.2$ | $1 / 24$ |  |
| 1.6 | $\leftrightarrow$ | 3.3 |
| 2.2 | $\leftrightarrow$ | 4.4 |

Table 2 - 048 input series VTM K factor selection guide

## Application Information (continued)

## OVP - Overvoltage Protection

The output overvoltage protection set point of the P036F048T12AL is factory preset for 56 V . If this threshold is exceeded the output shuts down and a restart sequence is initiated, also indicated by PC pulsing. If the condition that causes OVP is still present, the unit will again shut down. This cycle will be repeated until the fault condition is removed. The OVP set point may be set at the factory to meet unique high voltage requirements.

## PRM Output Power Versus VTM Output Power

As shown in Figure 17, the P036F048T12AL is rated to deliver 2.5 A maximum, when it is delivering an output voltage in the range from 26 V to 48 V , and 120 W , maximum, when delivering an output voltage in the range from 48 V to 55 V . When configuring a PRM for use with a specific VTM, refer to the appropriate VTM data sheet. The VTM input power can be calculated by dividing the VTM output power by the VTM efficiency (available from the VTM data sheet). The input power required by the VTM should not exceed the output power rating of the PRM.


Figure 17 - P036F048T12AL rating based on Factorized Bus voltage

The Factorized Bus voltage should not exceed an absolute limit of 55 V , including steady state, ripple and transient conditions. Exceeding this limit may cause the internal OVP set point to be exceeded.

## Parallel Considerations

The PR port is used to connect two PRMs in parallel to form a higher power array. When configuring arrays, PR port interconnection terminating impedance is 10 k to SG. See note Page 8 and refer to Application Note ANOO2. Additionally one PRM should be designated as the master while all other PRMs are set as slaves by shorting their SC pin to SG. The PC pins must be directly connected (no diodes) to assure a uniform start up sequence. Consult Vicor applications engineering for applications requiring more than two PRMs.

## Adjusting Current Limit

The current limit can be lowered by placing an external resistor between the $I_{L}$ and SG ports (see Figure 18 for resistor values). With the $I_{L}$ port open-circuit, the current limit is preset to be within the range specified in the output specifications table on Page 4.


Figure 18 - Calculated external resistor value for adjusting current limit, actual value may vary.

## Input Fuse Recommendations

A fuse should be incorporated at the input to the PRM, in series with the +In port. A fast acting fuse, NANO2 FUSE 451/453 Series 10 A 125 V , or equivalent, may be required to meet certain safety agency Conditions of Acceptability. Always ascertain and observe the safety, regulatory, or other agency specifications that apply to your specific application.

## Product Safety Considerations

If the input of the PRM is connected to SELV or ELV circuits, the output of the PRM can be considered SELV or ELV respectively.
If the input of the PRM is connected to a centralized DC power system where the working or float voltage is above SELV, but less than or equal to 75 V , the input and output voltage of the PRM should be classified as a TNV-2 circuit and spaced 1.3 mm from SELV circuitry or accessible conductive parts according to the requirements of UL60950-1, CSA 22.2 60950-1, EN60950-1, and IEC60950-1.

## Application Notes

For PRM and VI Chip application notes on soldering, board layout, and system design please click on the link below:
http://www.vicorpower.com/powerbench/applicationnotes

## Applications Assistance

Please contact Vicor Applications Engineering for assistance, 1-800-927-9474, or email at apps@vicorpower.com.

## Mechanical Drawings




## NOTES:

. DIMENSIONS
2. UNLESS OTHERWISE SPECIFIED, TOLERANCES ARE:
$. X /[. X X]=+/-0.25 /[.01] ; . \mathrm{XX} /[. \mathrm{XXX}]=+/-0.13 /[.005]$
3. PRODUCT MARKING ON TOP SURFACE

DXF and PDF files are available on vicorpower.com

Figure 19 - PRM J-Lead mechanical outline; Onboard mounting


NOTES:

1. DIMENSIONS ARE $\frac{\mathrm{mm}}{\mathrm{inch}}$
2. UNLESS OTHERWISE SPECIFIED, TOLERANCES ARE
$. X /[. X X]=+/-0.25 /[.01] ; . X X /[. X X X]=+/-0.13 /[.005]$
3. PRODUCT MARKING ON TOP SURFACE

DXF and PDF files are available on vicorpower.com

Figure 20 - PRM J-Lead PCB land layout information; Onboard mounting

## Mechanical Drawings (continued)



TOP VIEW (COMPONENT SIDE)



NOTES:


1. DIMENSIONS ARE $\frac{(\mathrm{mm})}{\mathrm{inch}}$.
2. UNLESS OTHERWISE SPECIFIED TOLERANCES ARE
$\mathrm{X} . \mathrm{X}[\mathrm{X} . \mathrm{XX}]= \pm 0.25[0.01] ; \mathrm{X} . \mathrm{XX}[\mathrm{X} . \mathrm{XXX}]= \pm 0.13[0.005]$
3. ROHS COMPLIANT PER CST-0001 LATEST REVISION

DXF and PDF files are available on vicorpower.com

Figure 21 - PRM Through-hole mechanical outline


NOTES:

1. DIMENSIONS ARE $\frac{(\mathrm{mm})}{\mathrm{inch}}$.
2. UNLESS OTHERWISE SPECIFIED TOLERANCES ARE: $X . X[X . X X]= \pm 0.25[0.01] ; X . X X[X . X X X]= \pm 0.13[0.005]$ 3. RoHS COMPLIANT PER CST-0001 LATEST REVISION

DXF and PDF files are available on vicorpower.com

Figure 22 - PRM Through-hole PCB layout information

## Configuration Options



Figure 23 - Hole location for push pin heat sink relative to VI Chip

## Vicor's comprehensive line of power solutions includes high density AC-DC and DC-DC modules and accessory components, fully configurable AC-DC and DC-DC power supplies, and complete custom power systems.

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VICOR'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS PRIOR WRITTEN APPROVAL OF THE CHIEF EXECUTIVE OFFICER AND GENERAL COUNSEL OF VICOR CORPORATION. As used herein, life support devices or systems are devices which (a) are intended for surgical implant into the body, or (b) support or sustain life and whose failure to perform when properly used in accordance with instructions for use provided in the labeling can be reasonably expected to result in a significant injury to the user. A critical component is any component in a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system or to affect its safety or effectiveness. Per Vicor Terms and Conditions of Sale, the user of Vicor products and components in life support applications assumes all risks of such use and indemnifies Vicor against all liability and damages.

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